# COMPLEX COEFFICIENT REPRESENTATION FOR IIR BILATERAL FILTER 

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11] reduce the computational order into $O(r)$. Constant time approaches $[12,13,14,15,16,17,18,19,20,21]$ is also proposed. The order of these filters is $O(1)$ per pixel. These approaches are extended to the optimization solver [22].

For acceleration of edge-preserving filtering, which does not limit bilateral filtering, several methods are proposed. Guided image filtering [4,23] is constructed from multiple box filtering, which is $O(1)$ filtering and is a FIR filter. Fast global image smoothing filtering [24], which is an acceleration method of weighted least square filtering [25], and is a global filter. Domain transform filtering [26] and recursive bilateral filtering [27] is a recursive implementation of infinite impulse response (IIR) based filtering.

The IIR based filtering is one of the fastest edge-preserving filters thanks to the cache efficiency and the capability of parallelization. The conventional IIR filters of the domain transform filtering and the recursive bilateral filtering require geodesic distance for the range kernel, while bilateral filtering requires Euclid distance. Also, these filters do not have separability for their kernels. The geodesic distance and nonseparable kernel make filters too sensitive to noise; thus, the response of the filtering output is not stable for noisy regions.

We proposed an IIR edge-preserving filtering with complex exponential coefficients for constructing a filter with Euclid (or Minkowski) distance based arbitrary range kernel and separable kernel. Also, we efficiently approximate bilateral filtering by using a raised-cosine function. We named this filter as IIR bilateral filtering. The main contribution of this paper is bridging the two types of filters, which are Euclid based filtering, e.g. bilateral filtering and geodesic based filtering, recursive bilateral filtering, in the IIR filtering domain.

## 2. COMPLEX COEFFICIENT FOR IIR FILTER

### 2.1. FIR and IIR Filter

1D FIR filtering over all signals can be represented as:

$$
\begin{equation*}
J_{p}=\sum_{q=1}^{N} W_{q, p} I_{q}, \tag{1}
\end{equation*}
$$

where $J_{p}, I_{q}$ is output and input pixel value, respectively, and $p, q$ are pixel location. $W_{q, p}$ is a weight between pixel $p$ and $q$ and $N$ is the number of pixels in the filtering kernel. In this case, $N$ is the same value as the number of input signals. All $J, W, I$ are real numbers $\in \mathbb{R}$ or complex numbers $\in \mathbb{C}$. Note that this equation is non-normalized version.

For IIR representation, we assume that the weight $W$ satisfies following assumptions:

$$
\begin{align*}
& W_{p, p}=1  \tag{2}\\
& W_{q, p}=W_{q, r} W_{r, p} \quad(q \leq r \leq p) . \tag{3}
\end{align*}
$$

With this assumption, Eq. (1) can be decomposed into leftside filtering $J_{p}^{L}$ and right-side filtering $J_{p}^{R}$ :

$$
\begin{align*}
J_{p} & =\sum_{q=1}^{p} W_{q, p} I_{q}+\sum_{q=p+1}^{N} W_{q, p} I_{q} \\
& =J_{p}^{L}+J_{p}^{R} . \tag{4}
\end{align*}
$$

We firstly focus the left-side filtering. The left-side filtering can be represented as first order, second order, ..., n-th order representation:

$$
\begin{align*}
J_{p}^{L} & =\sum_{q=1}^{p} W_{q, p} I_{q}=\sum_{q=1}^{p-1} W_{q, p} I_{q}+W_{p, p} I_{p} \\
& =W_{p-1, p} J_{p-1}^{L}+I_{p}  \tag{5}\\
& =W_{p-2, p-1} W_{p-1, p} J_{p-2}^{L}+W_{p-1, p} I_{p-1}+I_{p}  \tag{6}\\
& =\cdots \tag{7}
\end{align*}
$$

For the right-side case, we represent the filter as anti-causal way of the left-side filtering. After left-to-right and right-toleft filtering, both results are summed for obtaining the result.

In the case of bilateral filtering, we have a spatial kernel $S$ and a range kernel $R$. A weight $W_{q, p}=R_{q, p} S_{q, p}$ is defined as a multiplication of two Gaussian distributions $R_{q, p}=\exp \left(-\frac{1}{\sigma_{r}^{2}}\left|I_{q}-I_{p}\right|^{2}\right), S_{q, p}=\exp \left(\frac{-1}{\sigma_{s}^{2}}|q-p|^{2}\right)$. Note that the bilateral filter does not satisfy the assumptions of the weight (Eq. (2, 3)); thus, we extend this IIR representation for the bilateral filtering in Section 3.

### 2.2. 2D Filtering and Normalization

We can use separable filtering for 2D case under the assumption of Eq. $(2,3)$. We firstly filter horizontally and then vertically. Gaussian filtering is separable, and also complex exponential range kernel filtering is separable; thus, out separable filtering does not require approximation. Filtering with real number coefficients does not have the separable capability.

For image filtering, normalization is usually required. The normalized FIR filtering is:

$$
\begin{equation*}
J_{p}=\frac{\sum_{q=1}^{N} W_{q, p} I_{q}}{\sum_{q=1}^{N} W_{q, p}} \tag{8}
\end{equation*}
$$

The main difference of the upper and downer part of filtering is $I_{q}$ and 1. Thus, for efficient implementation, we use a vector of homogeneous coordinates $K_{q}=\left(I_{q}, 1\right)$ for filtering input instead of using the image $I_{q}$. Then, the division of elements in the smoothed result is performed:

$$
\begin{align*}
\bar{K}_{p} & =\sum_{q=1}^{N} W_{q, p} K_{q}  \tag{9}\\
J_{p} & =\bar{K}_{q}(0) / \bar{K}_{q}(1) \tag{10}
\end{align*}
$$

In this paper, we discuss only 1D filtering without anti-causal way of the left-side filtering due to the page limit.

### 2.3. Weight Definition

Under the assumption of Eq. (2, 3), the weight $W_{p, q}$ is defined by total product of the weight of adjoining pixels:

$$
\begin{equation*}
W_{q p}=W_{q, q+1} W_{q+1, q+2} \cdots W_{p-2, p-1} W_{p-1, p}=\prod_{j=q}^{p-1} W_{j, j+1} \tag{11}
\end{equation*}
$$

We present two representations of this weight in this section.

### 2.3.1. Real number weight for range kernel

We set the relation between the nearest pixels as a bilaterallike kernel, which satisfies the assumption of Eq. (2, 3):

$$
\begin{align*}
& R_{n, n+1}:=\exp \left(\frac{-\left|I_{n}-I_{n+1}\right|^{2}}{2 \sigma_{r}^{2}}\right)  \tag{12}\\
& S_{n, n+1}:=\exp \left(\frac{-1}{\sigma_{s}}\right) \tag{13}
\end{align*}
$$

In this case, $W_{p, q}$ is defined as:

$$
\begin{equation*}
W_{q, p}:=\exp \left(\frac{-|q-p|}{\sigma_{s}}\right) \exp \left(\frac{-\sum_{n=q}^{p-1}\left|I_{n}-I_{n+1}\right|^{2}}{2 \sigma_{r}^{2}}\right) \tag{14}
\end{equation*}
$$

The distance of the range kernel between $p$ and $q$ is total product of exponential functions so that the kernel representation expands total sum in a exponential function form the formula: $\prod_{n} \exp \left(x_{n}\right)=\exp \left(\sum_{n} x_{n}\right)$. The distance of Eq. (11) belongs to a geodesic distance. Note that the spatial kernel is not the Gaussian distribution, but the Laplace distribution.

This filter is easy to extend joint filtering [28, 29] by using guidance image $G$ instead of $I$ for range kernel computation.

$$
\begin{equation*}
W_{q, p}:=\exp \left(\frac{-|q-p|}{\sigma_{s}}\right) \exp \left(\frac{-\sum_{n=q}^{p-1}\left|G_{n}-G_{n+1}\right|^{2}}{2 \sigma_{r}^{2}}\right) \tag{15}
\end{equation*}
$$

### 2.3.2. Gaussian distribution for spatial kernel

The Gaussian distribution does not satisfy the assumption of Eq. (3), however, there are several approximation approaches in IIR filtering [30, 31, 32, 33, 34]. A recursive system of IIR filtering for general space-invariant filtering is usually represented as:

$$
\begin{equation*}
y_{p}=\sum_{l=0}^{m-1}\left(a_{l} x_{i-l}\right)-\sum_{k=1}^{m}\left(b_{k} y_{i-k}\right) \tag{16}
\end{equation*}
$$

where $y$ is output and $x$ is input. $m$ is the number of taps. $a, b$ are coefficients of the taps. We should set appropriate coefficients $a, b$ for the Gaussian distribution. For detail setting, please see Appendix, which presents implementations of the first/seconder order IIR Gaussian filter.

Plugging the IIR representation into the assumption of Eq. (3), we obtain IIR filtering with spatial and range kernel:

$$
\begin{equation*}
J_{p}^{L}=\sum_{l=0}^{m-1}\left(a_{l} R_{p, p-l} I_{p-l}\right)-\sum_{k=1}^{m}\left(b_{k} R_{p, p-k} J_{p-k}^{L}\right) \tag{17}
\end{equation*}
$$

When we set Gaussian distribution for spatial adjoint kernel, $S_{n, n+1}$, this equation is the same as the recursive bilateral filter [27].

### 2.3.3. Complex number weight for range kernel

Let we introduce the imaginary number $\mathrm{j}=\sqrt{-1}$ for coefficients to represent another kind of kernel. We use a complex exponential function for the range adjoint weight:

$$
\begin{equation*}
R_{p, p+1}:=\exp \left(\frac{-\mathrm{j}\left(I_{p}-I_{p+1}\right)}{\sigma_{r}}\right) . \tag{18}
\end{equation*}
$$

Note that we use just a subtraction, not an absolute or square difference, for intensity difference. Also, we do not extend $I$ (or $G$ ) to complex numbers; thus $I$ is real number.

Plugging in the Eq. (18) into Eq. (11), we can obtain the following range kernel:

$$
\begin{align*}
R_{q, p} & :=\exp \left(\sum_{n=q}^{p-1} \frac{-\mathrm{j}\left(I_{n}-I_{n+1}\right)}{\sigma_{r}}\right)=\exp \left(\frac{-\mathrm{j}\left(I_{q}-I_{p}\right)}{\sigma_{r}}\right)  \tag{19}\\
& =\cos \left(\frac{I_{q}-I_{p}}{\sigma_{r}}\right)-\mathrm{j} \sin \left(\frac{I_{q}-I_{p}}{\sigma_{r}}\right) . \tag{20}
\end{align*}
$$

Total product becomes a subtraction. Comparing with Eq. (13), this function does not use the geodesic distance. With this kernel, we can measure differences by using the Euclid distance with trigonometric functions (See Section 3.1).

## 3. EXTENSION FOR BILATERAL FILTERING

### 3.1. Fourier Series Decomposition for Arbitrary Kernel

Separating the real and imaginary part from the complex exponential kernel filtering in Eq. (17), we obtain result of filtering with trigonometric range weight, which are sin and cos

$$
\begin{equation*}
R_{q p}^{c}=\cos \left(\frac{I_{q}-I_{p}}{\sigma_{r}}\right), R_{q, p}^{s}=\sin \left(\frac{I_{q}-I_{p}}{\sigma_{r}}\right) . \tag{21}
\end{equation*}
$$

The theory of Fourier series decomposition, we can construct an arbitrary function $f$ whose argument is relative variable of $\sigma$ from the trigonometric function:

$$
\begin{equation*}
R_{q p}=f\left(\frac{x}{\sigma}\right)=\alpha_{0}+\sum_{n=1}^{\infty} \alpha_{n} \cos \left(\frac{2 n \pi}{\sigma} x\right)+\beta_{n} \sin \left(\frac{2 n \pi}{\sigma} x\right) \tag{22}
\end{equation*}
$$

where $\alpha_{n}, \beta_{n}$ are coefficients for Fourier series decomposition and $x=I_{q}-I_{p}$. sin and cos filtering result is obtained From the Eq. (eq:sincos). With the limited number of trigonometric functions, we can approximate an arbitrary function, e.g. Gaussian.

### 3.2. Raised-Cosine Approximation

For the Gaussian function, more effective approximation is presented. We use a raised-cosine function for the approximation introduced by the paper [18].

The raised-cosine in the range $T$ is represented as:

$$
\begin{equation*}
\cos ^{M}\left(\frac{x}{\sigma \sqrt{M}}\right) \quad(-T \leq x \leq T) . \tag{23}
\end{equation*}
$$



Fig. 1: Approximation of Gaussian ( $\sigma_{s}=60, M=4$ ).


Fig. 2: PSNR w.r.t $M$. $\sigma_{s}$ is 9 and $\sigma_{r}$ is from 20-100.
The function is converging to the Gaussian function with the limiting value of $M$ :

$$
\begin{equation*}
\lim _{M \rightarrow \infty} \cos ^{M}\left(\frac{x}{\sigma \sqrt{M}}\right)=\exp \left(-\frac{x^{2}}{2 \sigma^{2}}\right) . \tag{24}
\end{equation*}
$$

Applying the binomial theorem with Euler's formula $\cos \theta=$ $e^{\mathrm{j} \theta}+e^{-\mathrm{j} \theta}$, we can approximate the Gaussian function with the limited number of coefficients $M$ :

$$
\begin{align*}
R_{q p} & =\cos ^{M}\left(\frac{x}{\sigma}\right)=\sum_{m=0}^{M}\binom{M}{m} \exp \left(\mathrm{j}(2 m-M) \frac{x}{\sigma \sqrt{M}}\right)  \tag{25}\\
& \approx \exp \left(-\frac{x^{2}}{2 \sigma^{2}}\right), \tag{26}
\end{align*}
$$

where $x=I_{q}-I_{p}$.
Figure 1 shows the approximated Gaussian. The raisedcosine function can approximate the Gaussian function with fewer coefficients than the Fourier series decomposition.

## 4. EXPERIMENTAL RESULT

We evaluate the proposed method of IIR bilateral filtering (IBF) by using "Kodak Lossless True Color Image Suite" dataset. We use Intel Xeon X5690 3.47 GHz (dual-CPU24 thread). The code is written in C++ (Visual Studio 2013), and OpenMP is used for parallelization.

Figure 2 shows PSNR between the IBF and the ground truth. The ground truth is generated by the brute-force bilateral filter whose kernel radius is $3 \sigma_{s}$. The resulting PSNR is the average of 24 images in the dataset. When $\sigma_{r}$ is small, we need more coefficients. Also, if we have enough coefficients, increasing PSNR per the number of coefficients is few.

Figure 3 shows computational time w.r.t $M$. The computational cost is linearly related to $M$. In the small coefficient case, the IBF has real-time performance.

Figure 4 shows the results of IBF and recursive bilateral filtering (RBF) [27], which is given in Sec 2.2.1. RBF looks smoother than IBF with the same parameters. Table 1 shows


Fig. 3: Computational time w.r.t $M$.


Fig. 4: Filtering results of RBF (left) and IBF (right). The parameter is $\sigma_{c}=50, \sigma_{s}=50$.
the PSNR w.r.t various $\sigma_{r}, \sigma_{s}$. RBF is not an approximation of bilateral filtering, so that approximation accuracy is low when both $\sigma_{c}$ and $\sigma_{s}$ is large. On the contrary, IBF approximates well excepting for the small $\sigma_{r}$ case.

Figure 5 visualizes impulse responses of each filter. In the "Lenna" image overlaid dark cross lines, RBF, which uses geodesic distances, cannot travel the path of cross lines, while IBF, which uses Euclid distances, can pass. In the 13-th"Kodak" image of a high-frequency region, RBF cannot cover the region well due to non-separability and usage of geodesic distances, but the IBF can.

## 5. CONCLUSION

We introduce complex exponential coefficients for IIR filtering to represent the Euclid distance based filtering, e.g. bilateral filtering. We named this filter IIR bilateral filtering. Thanks to the raised-cosine function, we can approximate bilateral filtering with the fewer number of coefficients.

Limitation of IIR bilateral filtering is that loss of significant digits is critical when the order of IIR filtering is high. Even the second order approximation with small $\sigma_{r}$ makes the loss, so that we need robust IIR systems for this problem. In addition, this representation does not support color filtering; thus we will extend this work to color filtering by using [21].

## Appendix: IIR Gaussian Filter

In this section, we show coefficient of the first order (Eq. (27)) and second order (Eq. (28)) IIR Gaussian filtering:

$$
\begin{align*}
& y_{p}:=a_{0} x_{p}+y_{p-1}  \tag{27}\\
& y_{p}:=a_{0} x_{p}+a_{1} x_{p-1}-b_{1} y_{p-1}-b_{2} y_{p-2} \tag{28}
\end{align*}
$$

Table 1: PSNR of RBF (left) and IBF (right) w.r.t $\sigma_{s}$ and $\sigma_{c}$. The number of coefficients $\mathrm{M}=8$.

| V | or |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OS | ) | 10 | 30 | 50 | 70 | 90 |
|  | 3 | 39.21 | 35.41 | 35.02 | 35.19 | 35.53 |
|  | 9 | 35.95 | 30.64 | 30.10 | 30.61 | 31.52 |
|  | 15 | 35.00 | 29.12 | 28.46 | 29.06 | 30.17 |
|  | 21 | 34.52 | 28.25 | 27.45 | 28.05 | 29.23 |
|  | 27 | 34.22 | 27.65 | 26.74 | 27.30 | 28.47 |


|  | or |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\sigma S$ | V | 10 | 30 | 50 | 70 | 90 |
|  | 3 | 22.40 | 39.56 | 43.38 | 42.80 | 42.54 |
|  | 9 | 19.73 | 35.28 | 43.22 | 42.74 | 42.71 |
|  | 15 | 18.89 | 33.94 | 42.88 | 42.46 | 42.64 |
|  | 21 | 18.37 | 33.29 | 42.65 | 42.37 | 42.61 |
|  | 27 | 17.98 | 32.86 | 42.47 | 42.24 | 42.43 |



Fig. 5: Visualized kernel of RBF (left) and IBF (right). Each filtering parameter of $\sigma_{s} a n d \sigma_{c}$ is same. Note that the upper images are overlaid by dark cross lines. The images and kernels are cross dissolving with $1: 1$ ratio.

## First order Alvarez-Mazorra's IIR Gaussian filter

Alvarez-Mazorra's Gaussian approximation [31] is:

$$
\begin{align*}
a_{0} & =\frac{1+2 \lambda-\sqrt{(1+4 \lambda)}}{2 \lambda}  \tag{29}\\
\lambda & =q^{2} / 2 K  \tag{30}\\
q & =\sigma\left(1+\frac{0.3165 K+0.5695}{(K+0.7818)^{2}}\right), \tag{31}
\end{align*}
$$

where $K$ is the number of iterations. If we need more iteration, the filtering result becomes the iteration of bilateral filtering. We use $K=1$, in this paper.

## Second order Deriche's IIR Gaussian filter

The Deriche's second order Gaussian filter [30] is:

$$
\begin{align*}
a_{0} & =\left(1-c_{0}\right)^{2} /\left(1+3.390 c_{0} / \sigma_{s}-c_{1}\right)  \tag{32}\\
a_{1} & =\left(1.695 / \sigma_{s}-1\right) c_{0} a_{0}  \tag{33}\\
b_{1} & =-2 c_{0}  \tag{34}\\
b_{2} & =c_{1} \tag{35}
\end{align*}
$$

where $c_{0}=\exp \left(-1.695 / \sigma_{s}\right), c_{1}=\exp \left(-3.390 / \sigma_{s}\right)$.

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